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# Chung et al.

#### (54) FULL-DIGITAL CLOCK CORRECTION CIRCUIT AND METHOD THEREOF

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#### (56) **References Cited**

#### U.S. PATENT DOCUMENTS

5,614,855	A *	3/1997	Lee et al.	327/158
6,677,792	B2 *	1/2004	Kwak	327/158
6,765,421	B2 *	7/2004	Brox et al.	327/175
7,046,059	B2 *	5/2006	Kwak	327/158
7,057,431	B2 *	6/2006	Kwak	327/158
7,142,026	B2 *	11/2006	Kwak	327/158
7,250,801	B2 *	7/2007	Minzoni	327/175
7,671,647	B2 *	3/2010	Gomm et al.	327/158
7,750,703	B2 *	7/2010	Yun et al.	327/175
7,940,103	B2 *	5/2011	Satoh et al.	327/175
7,999,589	B2 *	8/2011	Lin	327/175
8,339,167	B2 *	12/2012	Gomm et al.	327/158

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2004/0066873	A1*	4/2004	Cho et al.	375/376
2006/0091921	A1*	5/2006	Park et al.	327/175
2007/0013423	A1*	1/2007	Lee	327/175
2007/0247203	A1*	10/2007	Cho et al.	327/175
2008/0001643	A1 $*$	1/2008	Lee	327/175
2008/0204099	A1*	8/2008	Hur	327/175
2010/0060334	A1*	3/2010	Abe et al.	327/158
2010/0117702	A1 $*$	5/2010	Jang et al.	327/175
2010/0207675	A1*	8/2010	Miyano	327/175
2010/0225372	A1*	9/2010	Satoh et al.	327/175
2010/0237917	A1*	9/2010	Monma	327/158
2011/0001527	A1*	1/2011	Lee	327/158

#### (Continued)

# OTHER PUBLICATIONS

Ching-Che Chung, Duo Sheng, and Sung-En Shen; A Wide-Range All-Digital Duty-Cycle Corrector with Output Clock Phase Alignment in 65 nm CMOS Technology; IEICE Electronics Express, vol. 8, No. 15, Aug. 10, 2011, pp. 1245-1251.

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#### (57) ABSTRACT

The present invention provides a full-digital clock duty cycle correction circuit and a method thereof. The circuit comprises a sampling unit, a duty cycle correcting module, and a phaselock module. The duty cycle correcting module produces a first clock signal according to an input clock signal. The phase-lock module produces a second clock signal according to the first clock signal and is used for aligning the positive edges of the clock signals. The duty cycle correcting module adjusts the pulse width of the first clock signal according to the clock signals. In addition, after the pulse width is adjusted, the positive edges of the clock signals are re-aligned. When the pulse width is not equal to zero, the pulse width is readjusted and the positive edges are re-aligned until the pulse widths of the clock signals are identical. Finally, the second clock signal is outputted and thus producing a clock signal having 50% duty cycle.

#### 8 Claims, 6 Drawing Sheets



# U.S. PATENT DOCUMENTS

2011/0074479	A1*	3/2011	Yun et al.	327/158
2011/0128059	A1*	6/2011	Kim et al.	327/175
2011/0204948	A1*	8/2011	Satoh et al.	327/175
2011/0248756	A1*	10/2011	Abe	327/158

2012/0124409	A1*	5/2012	Miyano	713/503
2012/0182054	A1*	7/2012	Sabapathy et al	327/172
2012/0268171	A1*	10/2012	Willey et al.	327/117

\* cited by examiner





Figure 2A





Figure 3



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Figure 4



Figure 5

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### FULL-DIGITAL CLOCK CORRECTION CIRCUIT AND METHOD THEREOF

#### FIELD OF THE INVENTION

The present invention relates generally to a duty cycle correction device and a method thereof, and particularly to a full-digital clock duty cycle correction circuit and a method thereof.

#### BACKGROUND OF THE INVENTION

Digital devices are prevalent nowadays. While spreading, exchanging, and processing digital information using digital devices, it is necessary to operate in coordination with clocks in circuits for processing digital signals or even the data attached in digital signals. For example, the central processing unit (CPU) in computer architecture should operate in coordination with the clock signal for driving the various  $_{20}$ digital circuits inside the CPU to operate in coordination with the clock signal for accessing data, processing data, or controlling hardware. In addition, in mobile devices, it is also required to use clocks for triggering transmission of digital information; synchronization in clocks is required before cor- 25 rect data access between digital devices. In the process of clock synchronization described above, the adopted technologies can be categorized into frequency division and frequency multiplication for giving, though different frequencies, harmonic and synchronized clocks, which would 30 facilitate coordination of digital data processing among different digital circuits.

In order to accelerate extraction of digital data, digital devices need to use the positive and negative edges of clock signals, which require the duty cycle of the clock signals to be 35 50%. Nonetheless, owing to imbalanced charge and discharge time via transistors and variations in process, temperature, and operating voltage, the duty cycle of clock signals in digital devices is not equal to 50%. Thereby, it is not possible to adopt the positive and negative edges of clock signals for 40 accelerating digital data extraction. Besides, errors may occur during data acquisition. For solving the problems described above, it is required to use a correction circuit for duty cycle to change the pulse width so that the duty cycle of the clock signals can recover to 50%.

As compared to an analog duty-cycle corrector (DCC), a full-digital DCC does not use the voltage control method, and thus it is less influenced by the leakage current of transistors especially in advanced CMOS processes. In addition, the all-digital DCC owns the advantages of fast lock-in time and 50 has a wide range of duty cycle correction. Thereby, general semiconductor digital devices, for example, memory devices, use extensively digital DCC to correct the duty cycle of clock signals. The digital DCC according to the prior art adopts a time to digital converter (TDC) to quantize the period of the 55 reference clock and then generates a half-cycle delay signal via a delay circuit for generating a clock signal with 50% duty cycle. Nevertheless, while adopting the time to digital converter architecture with a wide range of clock frequency, because the TDC accuracy has to be maintained and the 60 operating frequency has to be wide, the overall circuit area occupied by the digital DCC will increase significantly.

Furthermore, by using the TDC architecture, the half-cycle delay circuit has to be included, which makes the tuning accuracy of the digital DCC become twice the accuracy of the 65 TDC, and hence limiting the accuracy of the digital DCC and resulting in excessive error in duty cycle.

Accordingly, the present invention provides a full-digital clock correction circuit and a method thereof, which improve the problems using the TDC architecture according to the prior art. In addition, the accuracy in clock duty cycle correction is improved, and the frequency range and the duty cycle range of the supported clock signals are widened as well.

#### SUMMARY

An objective of the present invention is to provide a fulldigital clock duty cycle correction circuit and a method thereof, which improve duty-cycle correction accuracy and provide a wider frequency operating range. In addition, the present invention also supports a wider duty-cycle range of the input clock and reduces the chip area.

The present invention provides a full-digital clock duty cycle correction circuit, which comprises a sampling unit, a duty cycle correcting module, and a phase-lock module. The duty cycle correcting module is connected electrically to the sampling unit: the phase-lock module is connected electrically to the sampling unit and the duty cycle correcting module. The sampling unit receives an input clock signal and transmits it to the duty cycle correcting module for producing a first clock signal CLK\_X correspondingly. The phase-lock module produces a second clock signal CLK Y according to the first clock signal CLK\_X. Besides, the second clock signal CLK\_Y is the inverse of the first clock signal CLK\_X and is delayed in phase by a phase tuning unit. The phase-lock module detects the phase difference between the positive edges of the first and second clock signals CLK\_X, CLK\_Y selected by the sampling unit and aligns the positive edges of the first and second clock signals CLK\_X, CLK\_Y. The duty cycle correcting module compares the pulse width difference between the first and second clock signals CLK X, CLK Y for adjusting the pulse width of the first clock signal CLK\_X. When the pulse width of the first clock signal CLK\_X is changed, the pulse width of the second clock signal CLK\_Y will be changed accordingly. In addition, each time after the duty cycle correcting module adjusts the pulse width of the first clock signal CLK\_X, the phase-lock module needs to re-align the positive edges of the first and second clock signals CLK\_X, CLK\_Y. Then, when the pulse width difference between the first and second clock signals CLK\_X, CLK\_Y is not equal to zero, the duty cycle correcting module re-adjusts the pulse width of the first clock signal CLK\_X for the phaselock module to re-align the positive edges of the first and second clock signals CLK\_X, CLK\_Y until the first and second clock signals CLK\_X, CLK\_Y have identical pulse widths and the phase difference between the positive edges of the first and second clock signals CLK\_X, CLK\_Y is zero. Afterwards, the sampling unit selects an input clock signal CLK\_in and an output clock signal CLK\_out to the phase detecting unit for the phase-lock module to re-align the positive edges of the output clock signal CLK\_out and the input clock signal CLK\_in until the phase difference between the positive edges of the output clock signal CLK\_out and the input clock signal CLK\_in is zero. Thereby, the positive edge of the output clock signal CLK\_out is aligned with the input clock signal CLK\_in; the duty cycle of the output clock signal CLK\_out is also corrected to 50%.

After the phase-lock module aligns the positive edges of the first and second clock signals CLK\_X, CLK\_Y for the first time, if the duty cycle correcting module detects that the pulse width of the first clock signal CLK\_X is greater than the second clock signal CLK\_Y, namely, the duty cycle of the input clock signal is greater than 50%, the sampling unit selects the inversed input clock signal and transmits it to the 5

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duty cycle correcting module for ensuring that the duty cycle of the clock signal input to the duty cycle tuning unit is always less than 50%. At this moment, the output clock signal is changed to the inversed second clock signal CLK\_Y.

#### BRIEF DESCRIPTION OF THE DRAWINGS

FIG. 1 shows a circuit diagram according to an embodiment of the present invention;

FIG. 2A and FIG. 2B show waveforms according to an 10 embodiment of the present invention;

FIG. 3 shows a circuit diagram of the coarse-tuning phase detecting circuit according to the present invention;

FIG. 4 shows a circuit diagram of the fine-tuning phase detecting circuit according to the present invention; and

FIG. 5 shows a flowchart according an embodiment of the present invention.

#### DETAILED DESCRIPTION

In order to make the structure and characteristics as well as the effectiveness of the present invention to be further understood and recognized, the detailed description of the present invention is provided as follows along with embodiments and accompanying figures.

FIG. 1 shows a circuit diagram according to an embodiment of the present invention. As shown in the figure, the full-digital clock duty cycle correction circuit 10 according to the present invention comprises a sampling unit 12, a duty cycle correcting module 14, and a phase-lock module 16. In 30 addition, the sampling unit 12 further comprises a first multiplexer 122, a second multiplexer 124, a third multiplexer 126, a fourth multiplexer 128, a first inverter 130, and a second inverter 132. The duty cycle correction module 14 further comprises a duty cycle detecting unit 142, a duty cycle 35 tuning unit 144, and a duty cycle control unit 146. Besides, the phase-lock module 16 further comprises a phase detecting unit 162, a phase tuning unit 164, a phase control unit 166, and a third inverter 168.

The first multiplexer 122 is connected electrically to the 40 input clock signal CLK\_in. The first inverter 130 is connected electrically to the third multiplexer 126, the duty cycle tuning unit 144, and the duty cycle control unit 146. The second multiplexer 124 is connected electrically to the fourth multiplexer 128, the second inverter 132, the duty cycle detecting 45 unit 142, the duty cycle control unit 146, the phase tuning unit 164, the input of the fourth multiplexer 128, and the output clock signal CLK\_out. The third multiplexer 126 is connected electrically to the duty cycle detecting unit 142, the duty cycle tuning unit 144, the phase detecting unit 162, the 50 phase tuning unit 164, and the phase control unit 166. The fourth multiplexer 128 is connected electrically to the duty cycle detecting unit 142, the phase detecting unit 162, the phase tuning unit 164, and the phase control unit 166.

The sampling unit 12 is connected electrically to the phase- 55 lock module 16 and the duty cycle correcting module 14. The sampling unit 12 receives an input clock signal CLK\_in and transmits it to the duty cycle correcting module 14 for producing a first clock signal CLK\_X correspondingly. The phase-lock module 16 receives the first clock signal CLK\_X 60 and produces correspondingly a second clock signal CLK\_Y, which is the inverse of the first clock signal CLK\_X and phase delayed by the phase tuning unit 164.

The duty cycle correcting module 14 detects the pulse width difference between the first and second clock signals 65 CLK\_X, CLK\_Y and adjusts the pulse width of the first clock signal CLK\_X according to the pulse-width difference. The

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duty cycle detecting unit 142 produces a duty cycle detecting signal D det according to the difference between the pulse widths of the first and second clock signals CLK\_X, CLK\_Y. The duty cycle control unit 146 produces a duty cycle control code D\_code according to the duty cycle detecting signal D\_det and transmits the duty cycle control code D\_code to the duty cycle correcting unit 144. The duty cycle tuning unit 144 receives the first clock signal CLK\_X, and adjusts the pulse width of the first clock signal CLK\_X according to the duty cycle control code D\_code. When the pulse width of the first clock signal CLK X is changed, the pulse width of the second clock signal CLK\_Y is changed accordingly. In addition, after the duty cycle control unit 146 adjusts the pulse widths described above, the duty cycle control unit 146 locks the duty cycle correcting unit 144. The duty cycle control unit 146 will control the first and second multiplexers 122, 124 via a duty cycle select signal duty\_select according to whether the duty cycle of the input signal is greater or less than 50%.

The phase-lock module 16 detects and compensates the 20 phase difference between the positive edges of the first and second clock signals CLK\_X, CLK\_Y for aligning them. The phase detecting unit 162 detects the phase difference of positive edges for producing a phase detecting signal P\_det and transmitting it to the phase control unit 166. The phase control unit 166 produces a phase control code P code according to the phase detecting signal P\_det and transmits it to the phase tuning unit 164. The phase tuning unit 164 receives the inverse signal of the first clock signal CLK\_X produced by the third inverter 168, and tunes and aligns the phases of the positive edges of the first and second clock signals CLK\_X, CLK\_Y according to the phase control code P\_code. Besides, after the phases of the first and second clock signals CLK\_X, CLK\_Y are aligned, the phase control unit 166 further locks the phase tuning unit 164. After the duty cycle is corrected, the phase control unit 166 further controls the third and fourth multiplexer 126, 128 via a phase select signal phase\_select for selecting the input clock signal CLK\_in and the output clock signal CLK\_out to the phase detecting unit 162 for the phase-lock module 16 to re-align the positive edges of the output and input clock signals CLK\_out, CLK\_in until the phase difference between the positive edges of the output and input clock signals CLK\_out, CLK\_in is zero. Then, the positive edge of the output clock signal CLK\_out is aligned with the input clock signal CLK\_in, and the duty cycle thereof is corrected back to 50%.

Every time the duty cycle correcting module 14 adjusts the pulse width of the first clock signal CLK\_X, the phase-lock module 16 needs to re-align the positive edges of the first and second clock signals CLK\_X, CLK\_Y. When the pulse-width difference is not equal to zero, the duty cycle correcting module 14 re-adjusts the pulse width of the first clock signal CLK X for the phase-lock module 16 to re-align the positive edges of the first and second clock signals CLK\_X, CLK\_Y until the first and second clock signals CLK\_X, CLK\_Y have identical pulse widths and the phase difference between the positive edges of the first and second clock signals CLK\_X, CLK\_Y is zero. Then, when the duty cycle of the input clock signal CLK\_in is less than 50%, the second clock signal CLK\_Y is outputted as the output clock CLK\_out. On the other hand, when the duty cycle of the input clock signal CLK\_in is greater than 50%, the output is changed to the inverse of the second clock signal CLK\_Y. After the fulldigital clock duty cycle correction circuit 10 is locked, the duty-cycle of the output clock signal CLK\_out will be 50%.

FIGS. 2A and 2B show waveforms according to an embodiment of the present invention. As shown in FIG. 2A, after system reset, the duty cycle select signal duty\_select outputted by the duty cycle correcting module 14 is zero, namely, a low signal level. The phase select signal phase\_select outputted by the phase-lock module 16 is also zero, a low signal level. The sampling unit 12 receives the input clock signal CLK\_in and transmits it to the duty cycle correcting module 14, which, then, outputs the first clock signal CLK X. Next, the first clock signal CLK X is inverted and input to the phase-lock module 16, which outputs the second clock signal CLK\_Y. The phase detecting unit 162 of the phase-lock module 16 detects the phase difference between the first and second clock signals CLK\_X, CLK\_Y and outputs the corresponding detecting signal P\_det to the phase control unit 166. The phase control unit 166 produces the corresponding phase control code P\_code according to the detecting signal P\_det and transmits it to the phase tuning unit 164 for compensating the phase difference between the first and second clock signals CLK\_X, CLK\_Y. After the phase difference between the first and second clock signals CLK\_X, CLK\_Y is eliminated, the phase-lock module 16 is locked. 20 After the phase-lock module 16 is locked, the positive edges of the first and second clock signals CLK\_X, CLK\_Y are aligned. Besides, because the first and second clock signals CLK\_X, CLK\_Y are complementary signals, when the period of the input clock signal CLK\_in is T and initial pulse 25 widths of the first and second clock signals CLK\_X, CLK\_Y are assumed to be A and B, their duty cycles will be (A/T)\*100% and (B/T)\*100%, respectively, where T=A+B.

As shown in FIG. 2B, after the phase-lock module 16 is locked, the duty cycle correcting module 14 starts to compen- 30 sate the duty-cycle error of the output clock signal CLK\_out, namely, adjusting the pulse width of the clock signal. The duty cycle detecting unit 142 of the duty cycle correcting module 14 detects the negative edges of the positive-edgealigned first and second clock signals CLK\_X, CLK\_Y. In 35 548. other words, as the phases of the positive edges of the first and second clock signals CLK\_X, CLK\_Y are aligned, the phases of the negative edges thereof are compared for giving the pulse-width difference between the first and second clock signals CLK\_X, CLK\_Y and producing the corresponding 40 detect signal D\_det and transmitting it to the duty cycle control unit 146. The duty cycle control unit 146 produces the corresponding duty cycle control code D\_code according to the detect signal D\_det and transmits it to the duty cycle tuning unit 144 for widening the pulse width of the first clock 45 signal CLK\_X. According to the present embodiment, it is assumed that the duty cycle of the input clock signal CLK\_in is less than 50%. Nonetheless, the present invention is not limited to the embodiment.

Every time the duty cycle correcting module **14** widens the 50 pulse width of the first clock signal CLK\_X, then, at the next cycle, because the pulse width of the previous cycle is tuned, the phase of the positive edge of the second clock signal CLK\_Y lags that of the first clock signal CLK\_X. Thereby, at the second cycle, the phase control code P\_code is adjusted 55 for re-aligning the positive edges of the first and second clock signals CLK\_X, CLK\_Y. Accordingly, the phases of the positive and negative edges of the first and second clock signals CLK\_X, CLK\_Y are corrected repeatedly until the pulse widths of the first and second clock signals CLK\_X, CLK\_Y are corrected repeatedly until the pulse widths of the first and second clock signals CLK\_X, CLK\_Y are lock signals CLK\_X, CLK\_Y for are identical. Then the duty cycle correcting module **14** is locked.

The pulse width of the first clock signal CLK\_X increases by a shift value  $\Delta E$ , which is equal to (B–A)/2. Because the period of the input clock signal is T, the pulse widths of the 65 first and second clock signals CLK\_X, CLK\_Y, finally, become A+ $\Delta E$ =A+(B–A)/2=(A+B)/2=T/2, which means

that the duty cycles of the first and second clock signals CLK X, CLK Y are corrected to (T/2)/T\*100%=50%.

Furthermore, when the duty cycle correcting module **14** is locked, the signal level of the phase select signal phase\_select outputted by the phase-lock module is driven to high for switching the input signal of the phase detecting unit **162** of the phase-lock module **16** to the input clock signal CLK\_in and the output clock signal CLK\_out. Then, the phase-lock module **16** can further compensate the phase difference according to the input and output clock signals CLK\_in, CLK\_out for aligning the positive edges thereof.

After the phase-lock module 16 aligns the positive edges of the first and second clock signals CLK\_X, CLK\_Y for the first time, if the duty cycle correcting module 14 detects the pulse width of the first clock signal CLK\_X greater than that of the second clock signal CLK\_Y, namely, the duty cycle of the input clock signal CLK\_in is greater than 50%, the signal level of the duty cycle select signal duty\_select outputted by the duty cycle correcting module 14 is high for inverting the input clock signal CLK\_in, and thus ensuring that the duty cycle of the input signal to the duty cycle tuning unit 144 is always less than 50%. At this moment, the output clock signal is changed to the inverted second clock signal CLK\_Y.

FIG. 3 and FIG. 4 show circuit diagram of phase detecting units according to the present invention. As shown in the figures, the phase detecting unit according to the present invention comprises a coarse-tuning phase detecting circuit 51 and a fine-tuning phase detecting circuit 52. The coarsetuning phase detecting circuit 51 comprises an inverter 511, a buffer 512, a first flip-flop 514, a second flip-flop 516, and a logic controller 518; the fine-tuning phase detecting circuit 52 comprises a plurality of inverters 522, 524, 526, 528, 530, 532, 534, a plurality of transistors 536, 538, 540, 542, 544, a first transmission gate 546, and a second transmission gate 548.

The first and second flip-flops **514**, **516** of the coarsetuning phase detecting circuit **51** are D flip-flops. The buffer **512** is connected to the clock signal Y and the second flip-flop **516** via the inverter **511**. The first flip-flop **514** is connected to the clock signal X, and to the clock signal Y via the inverter **511**. Besides, the outputs Q1, Q2 of the first and second flip-flops **514**, **516** are led to the logic controller **518** so that the logic controller **518** can output the detecting signals, such as phase-leading signal UP\_1, phase-lagging signal DOWN\_1, and phase-lock signal LOCK.

The inverter 522 of the fine-tuning phase detecting circuit 52 receives the clock signal X and is connected to the gate of the transistor 540. The inverter 524 receives the clock signal Y and is connected to the gates of the transistors 536, 538. The inverter 526 receives the clock signal Y and is connected to the first and second transmission gates 546, 548. The inverter 528 outputs the phase-leading signal UP\_2. The inverter 530 is connected to the inverter 528, the gate of the transistor 542, and the first transmission gate 546. The inverter 532 outputs the phase-lagging signal DOWN\_2. The inverter 534 is connected to the inverter 532, the gate of the transistor 544, and the second transmission gate 548. The transistor 536 is further coupled to the ground and the transistors 538, 540. The transistor 542 is further coupled to the transistor 538 and the first transmission gate 546. The transistor 544 is further connected to the transistor 540 and the second transmission gate 548. The first and second transmission gates 546, 548 are connected to a voltage source  $V_{DD}$ , respectively.

Moreover, when the phase difference between the clock signals X, Y is very large, the coarse-tuning phase detecting circuit **51** is used for judging the situation of phase leading and lagging. When the phase difference between the clock

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signals X, Y is smaller than the detectable range of the coarsetuning phase detecting circuit **51**, the fine-tuning phase detecting circuit **52** is used for detecting the extremely small phase difference between the clock signals X, Y.

The coarse-tuning and fine-tuning phase detecting circuits 5 51, 52 according to the embodiment described above are used as the duty cycle detecting unit 142 in FIG. 1. The coarsetuning and fine-tuning phase detecting circuits 51, 52 described above detect the negative-edge phases for detecting the pulse widths of the first and second clock signals CLK\_X, 10 CLK\_Y in FIG. 1. In addition, by eliminating the inverter 511 in the coarse-tuning phase detecting circuit 51 and the inverters 522, 524, 526 in the fine-tuning phase detecting circuit 52, they can be used as the phase detecting unit 162 in FIG. 1.

FIG. **5** shows a flowchart according an embodiment of the 15 present invention. As shown in the figure, the full-digital clock duty cycle correction method according to the present invention comprises steps of:

Step S100: Receiving an input clock signal for producing a first clock signal and a second clock signal correspond- 20 ingly;

- Step S110: Detecting the phase difference between the positive edges of the first and second clock signals for judging if the positive edges of the first and second clock signals is aligned;
- Step S120: Aligning the positive-edge phases of the first and second clock signals according to the positive-edge phase difference;
- Step S130: Judging if the duty cycle of the first clock signal is greater than 50%;

Step S140: Inverting the first clock signal;

- Step S150: Comparing the pulse-width difference between the first and second clock signals for judging if the pulse widths of the first and second clock signals are identical;
- Step S160: Adjusting the pulse width of the first clock signal 35 according to the pulse-width difference;
- Step S170: Detecting a positive-edge phase differencebetween the first and second clock signals and aligning thepositive edges of the first and second clock signals accord-ing to the positive-edge difference; and40

Step S180: Selecting and outputting the output clock signal. In the step S100, the sampling unit 12 receives an input clock signal and inputs it to the duty cycle correcting module 14 for producing a first clock signal correspondingly and inputting it to the phase-lock module 16, which, then, pro- 45 duces a second clock signal correspondingly. The second clock signal is the inverted first clock signal. In the step S110, the phase-lock module detects the phase difference between the first and second clock signals sampled by the sampling unit 12 for judging if the first and second clock signals are 50 aligned. If the positive edges are not aligned, the step S120 is executed; if aligned, the step S130 is executed. In the step S120, the phase-lock module 16 compensates the measured phase difference between positive edges for aligning the positive edges of the first and second clock signals as well as 55 delaying the positive edge of the second clock signal until the positive edges of the first and second clock signals are aligned. The description of the present embodiment is as shown in FIG. 2A.

In the step S130, the positive-edge-aligned first and second 60 clock signals are received and transmitted to a duty cycle correcting module 14 for detecting the pulse widths of the first and second clock signals, which is used for judging if the duty cycle is greater than 50%. When the duty cycle of the first clock signal is greater than 50%, the step S140 is executed; 65 otherwise, the step S150 is executed. In the step S140, because the duty cycle of the first clock signal is greater than

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50%, the duty cycle correcting module 14 produces a duty cycle select signal duty select correspondingly for controlling the sampling unit 12 to receive the inverted clock signals, namely, the inverted input clock signal CLK\_in, the first clock signal CLK\_X, and the second clock signal CLK\_Y. Next, the step S120 is executed subsequently. In the step S150, the duty cycle correcting module 14 compares the difference of pulse widths between the first and second clock signals for judging if their pulse widths are equal. If their pulse widths are not equal, the step S160 is executed; otherwise, the step S180 is executed. In the step S170, after the pulse width of the first clock signal is adjusted, it is required to switch to the phaselock module 16 for re-detecting the positive-edge phase difference between the first and second clock signals, and hence re-aligning the positive edges of the first and second clock signals. Then the step S150 is executed for re-judging if the pulse widths of the first and second clock signals are equal until they have identical pulse widths. In the step S180, after the phase-lock module 16 aligns the first and second clock signals having identical pulse widths, it controls the sampling unit 12 to select the second clock signal and the inverted second clock signal. By means of the duty cycle select signal duty\_select, the sampling unit 12 is controlled to output the output clock signal.

Finally, the present invention provides a full-digital clock duty cycle correction circuit and a method thereof. The present invention conforms to the legal requirements owing to its novelty, nonobviousness, and utility. However, the foregoing description is only embodiments of the present invention, not used to limit the scope and range of the present invention. Those equivalent changes or modifications made according to the shape, structure, feature, or spirit described in the claims of the present invention are included in the appended claims of the present invention.

The invention claimed is:

- **1**. A full-digital clock duty cycle correction circuit, comprising:
- a sampling unit, receiving and sampling an input clock signal;
- a duty cycle correcting module, connected electrically to said sampling unit, and producing a first clock signal according to said input clock signal;
- a phase-lock module, connected electrically to said sampling unit and said duty cycle correcting module, inverting said first clock signal and producing a second clock signal, said sampling unit transmitting said first clock signal and said second clock signal to said phase-lock module, said phase-lock module detecting a positiveedge phase difference between said first clock signal and said second clock signal for aligning the positive edges of said first clock signal and said second clock signal, said sampling unit transmitting said positive-edgealigned first clock signal and said positive-edge-aligned second clock signal to said duty cycle correcting module, said duty cycle correcting module comparing a pulse width difference between said first clock signal and said second clock signal for adjusting the pulse widths of said first clock signal and said second clock signal until the pulse widths are identical, and outputting said second clock signal as a third clock signal.
- **2**. The full-digital clock duty cycle correction circuit of claim **1**, wherein said phase-lock module comprises:
  - a phase detecting unit, detecting the positive-edge phase difference between said first clock signal and said second clock signal to produce a phase detecting signal;
  - a phase tuning unit, receiving said first clock signal and said second clock signal; and

a phase control unit, connected electrically to said phase tuning unit and said phase detecting unit, producing a phase control code according to said phase detecting signal and transmitting said phase control code to said phase tuning unit for driving said phase tuning unit to aligning the positive edges of said first clock signal and said second clock signal, and said phase control unit locking said phase tuning unit after aligning the phases of said first clock signal.

**3**. The full-digital clock duty cycle correction circuit of 10 claim **1**, wherein said duty cycle correcting module comprises:

- a duty cycle detecting unit, producing a duty cycle detecting signal according to the pulse-width difference between said first clock signal and said second clock 15 signal;
- a duty cycle tuning unit, receiving said first clock signal and said second clock signal; and
- a duty cycle control unit, connected electrically to said duty cycle detecting unit, producing a duty cycle control code 20 according to said duty cycle detecting signal and transmitting said duty cycle control code to said duty cycle correcting unit for adjusting the pulse widths of said first clock signal and said second clock signal, and said duty cycle control unit locking said duty cycle correcting unit 25 after adjusting the duty cycle of said first clock signal and said second clock signal.

**4**. A full-digital clock duty cycle correction method, comprising steps of:

- receiving an input clock signal, and sampling for produc- 30 ing a first clock signal;
- inverting said first clock signal and giving a second clock signal;
- transmitting said first clock signal and said second clock signal to a phase-lock module for detecting a positive-35 edge phase difference between said first clock signal and said second clock signal, and aligning the positive edges of said first clock signal and said second clock signal according to the positive-edge phase difference;

- transmitting said positive-edge-aligned first clock signal and said positive-edge-aligned second clock signal to a duty cycle correcting module, and comparing the pulsewidth difference between said first clock signal and said second clock signal for adjusting the pulse widths of said first clock signal and said second clock signal; and
- comparing and outputting said second clock signal as a full-digital clock signal having 50% duty cycle when the pulse-width difference between said first clock signal and said second clock signal is zero.

**5**. The full-digital clock duty cycle correction method of claim **4**, wherein said step of detecting the positive-edge phase difference between said first clock signal and said second clock signal is detecting the positive-edge phase difference to produces a phase detecting signal.

**6**. The full-digital clock duty cycle correction method of claim **5**, wherein said step of aligning the positive edges of said first clock signal and said second clock signal according to the positive-edge phase difference is producing a phase control code according to said phase detecting signal and transmitting said phase control code to said phase-lock module for compensating the positive-edge phase difference.

7. The full-digital clock duty cycle correction method of claim 4, wherein said step of comparing the pulse-width difference between said first clock signal and said second clock signal is producing a duty cycle detecting signal according to the pulse-width difference between said first clock signal and said second clock signal.

**8**. The full-digital clock duty cycle correction method of claim **7**, wherein said step of adjusting the pulse widths of said first clock signal and second clock signal is producing a duty cycle control code according to the duty cycle detecting signal and transmitting said duty cycle control code to said duty cycle correcting module for adjusting the pulse widths of said first clock signal and second clock signal.

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